# SMDP-Based Radio Resource Allocation Scheme in Software-Defined Internet of Things Networks

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Abstract—With rapid development of the Internet of Things (IoT), various machine-to-machine (M2M) communications technologies have emerged in recent years to provide ubiquitous wireless connections for a massive number of IoT devices. This poses significant challenges to network control and management of large-scale IoT networks. Software-defined networking (SDN) is considered a promising technology to streamline network management due to dynamic reconfigurable network elements. Thus, the integration of SDN and IoT provides a potentially feasible solution to strengthening management and control capabilities of the IoT network. Benefit from SDN technology, resource utilization in the IoT network can be further enhanced. In this paper, we first propose a software-defined network architecture for IoT. Then, the resource allocation problem in the proposed SDN-based IoT network is investigated. The optimal problem of maximizing the expected average rewards of the network is formulated as a semi-Markov decision process (SMDP). The optimal solution is obtained through solving the SMDP problem using a relative value iteration algorithm. Simulation results demonstrate that the proposed resource allocation scheme is able to improve the system rewards compared with other comparative resource allocation schemes.

Keywords—IoT; SDN; semi-Markov decision process (SMDP).

#### I. INTRODUCTION

The advent of the Internet of Things (IoT) has inspired a large variety of new applications that can provide ubiquitous services to make existing industrial systems and peoples life more intelligent, e.g., industrial automation, smart grids, intelligent transportation systems, smart healthcare, smart home, etc [1]. However, with the rapid development of IoT technology, the management and control of IoT networks is met with considerable challenges. It is expected that there will be 100-fold or more machine-to-machine (M2M) connections in

the emerging fifth generation (5G) era [2], which requires the network not only to support the massive M2M connections, but also to be optimized with the aid of big data analytics [3]. Meanwhile, the massive number of connections by M2M devices will share the same network with human-to-human (H2H) users. The M2M connections often have different requirements on the network due to the unique characteristics of M2M communications [4] [5].

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Furthermore, various M2M communications technologies have emerged in recent years, although none of them can meet all the requirements of M2M connections. At present, M2M communications can be classified into two categories, i.e., cellular M2M solutions, and capillary M2M solutions [6]. To make cellular networks more suited for M2M communications, a great deal of efforts have been made in third generation partnership project (3GPP) recent releases by defining longterm evolution for machine-type communications (LTE-MTC), which aims for low power consumption and moderate costs [7]. Moreover, some clean-slate cellular M2M solutions are also developed to provide long terminal battery life and extended coverage, such as narrowband IoT (NB-IoT) [8]. On the other hand, capillary M2M solutions have been widely used in IoT applications for many years, which provide wireless connections for devices with low power consumption, such as ZigBee, Wi-Fi, etc. In addition, low power wide area (LPWA) technology as another promising solution has been proposed to meet the enhanced coverage requirement of some M2M application scenarios such as rural areas, which provides a long-range connectivity up to several kilometers, such as LoRa [9]. As a consequence, all the techniques and solutions for M2M communications may coexist and work together to support various IoT applications in the future. Thus, the heterogeneity of M2M connectivity also brings about significant challenges to dynamically optimize radio resource allocation and management in the IoT network.

To address these challenges, the software-defined networking (SDN) is considered as a promising candidate technology for simplifying network management due to its programmable and centralized network control [10]. SDN was originally proposed and designed for computer networks to decouple control decisions from forwarding devices such as switches and routers [11]. Many previous efforts have been made to migrate the concept of SDN to wireless networks, especially softwaredefined mobile networks (SDMN), such as Openradio [12], SoftRAN [13], SoftCell [14], CROWD [15], MobileFlow [16], SoftNet [17], SERVICE [18], etc. Moreover, the integration of SDN and IoT has also been attempted [19]. Sensor OpenFlow

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is proposed as a software-defined wireless sensor network (WSN) architecture to tackle the inherent problems of WSN, i.e., rigidity to policy changes and difficulty to manage [20]. In [21], the authors propose a SDN controller design for the IoT network, which enables centralized flow scheduling based on network calculus model. In [22], SDN-WISE is designed as a stateful SDN solution for the WSN, and the performance of SDN-WISE is also evaluated based on a prototype implementation. UbiFlow adopts a distributed hashing based overlay structure for flow scheduling and mobility management in a software-defined IoT system [23].

To the best of our knowledge, the research on integrating SDN and IoT networks is still at its infancy. Many problems remain open, e.g., network architecture, protocols, controller design, standardization, etc. In this paper, we first propose a software-defined network architecture for IoT, which is based on the 3GPP proposed architecture for machine type communications (MTC). Then, a novel radio resource allocation scheme is proposed for the software-defined IoT network. It is noted that issues relating to radio resource allocation in SDN-based IoT networks are rarely studied in the literature. We formulate the radio resource allocation problem in the SDN-based IoT network as a semi-Markov decision process (SMDP) [25], which provides a mathematical framework for modeling a wide range of optimization problems [26] [27] [28]. From a global network point of view, the SDN controller can make optimal decisions to maximize the expected average reward of the network. The optimal radio resource allocation policy is obtained through solving the SMDP problem using the relative value iteration algorithm [25]. To verify the performance of our proposed model, a simulation-based study is performed by comparing the performances of our model and other reference algorithms. Numerical results will be presented to demonstrate that our approach is able to enhance the overall resources utilization resulting in improved performance for the SDNbased IoT network.

The remainder of this paper is organized as follows. In Section II, the system model is described, and the resource allocation problem is formulated as an SMDP process. The solution to the SMDP problem is attempted in Section III. Then, Section IV provides numerical results as well as a performance analysis of our proposed scheme. Finally, concluding remarks are drawn in Section V.

### II. SYSTEM MODEL AND PROBLEM FORMULATION

In this section, we first describe a software-defined heterogeneous network architecture for IoT. Then, our proposed optimal radio resource allocation scheme for this SDN-based IoT network architecture is presented, which is formulated as an SMDP problem. Next, we describe the system states of our proposed model, and the actions that can be taken in each state. The system reward model is also described, which plays a significant role in decision making by the SDN controller.

## A. System Model

We consider an SDN-based IoT network as illustrated in Fig. 1, which basically consists of an eNB, N MTC gateways



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Fig. 1. Illustration of the system model.

(MTCGs), and a local SDN controller. As a new network element, the SDN controller is the control plan of the overall network. Most of the control logics are implemented in the controller, or rest in management applications that interact with the SDN controller via the northbound application program interface (API). The SDN controller has two essential functions. Firstly, it collects the overall status information of the eNB and each MTCG, and offers a global network view to the management applications. Secondly, the SDN controller makes decisions and controls the eNB and MTCGs dynamically based on the obtained network status, which can be implemented by programming through the southbound API, e.g., extensions to the OpenFlow protocol [11].

The eNB and MTCGs are responsible for providing radio access for MTC devices, and the coverage areas of the eNB and MTCGs overlap in the geographical region of the SDN-based IoT network. Thus, M MTC devices (MTCDs) dispersed in the serving area can access to the core network through either the eNB or relaying via an MTCG. The direct connection between an MTCD and the eNB is dubbed the MTCD-to-eNB link, while the MTCD-to-MTCG link describes the wireless link between an MTCD and an MTCG. We assume that the MTCDto-eNB links and MTCD-to-MTCG links adopt the same NB-IoT M2M communications technique, which is considered as the most potential candidate M2M technology for the future fifth generation (5G) wireless network. Compared to existing cellular techniques, NB-IoT provides limited transmission rates and extended coverage. Each MTCD-to-MTCG and MTCDto-eNB link can support  $l_m$  transmission modes with different transmission rates  $C_m(\cdot)$  in accordance with the signal-tonoise ratio (SNR), which indicates the quality of the channel. The SNR range can be split into  $l_m$  non-overlapping intervals with boundary points denoted by  $\{\{\theta_{m,i}\}_{i=1}^{l_m+1}\},\$ where  $\theta_{m,l_m+1} = \infty$ . When SNR  $\xi \in [\theta_{m,i}, \theta_{m,i+1})$ , the transmission mode i is applied and its transmission rate  $C_m(\xi)$ is equal to  $\mu_{m,i}$ , where  $i \in \{1, \ldots, l_m\}$ . Suppose each MTCG, as well as the eNB, owns K radio resources (termed access resources (ARs)), which are assigned to the MTCD-to-MTCG

and MTCD-to-eNB links. Each MTCD-to-MTCG link and MTCD-to-eNB link can consume u ARs of the connected MTCG and eNB, where  $u \in \{1, \ldots, U\}$  and  $U \leq K$ . Moreover, data packets transmitted from an MTCD follow a Poisson distribution with the mean rate of  $\lambda_m$ , and the MTC packet size follows an exponential distribution with a mean size of  $\rho_m$ . Therefore, the total arrival rate of the MTC data packets to the network is equal to  $M\lambda_m$ .

When an MTCD is connected to an MTCG, the MTCG performs as a one-hop relay to forward data packets to the eNB via the MTCG-to-eNB link. It is assumed that the MTCG-toeNB link employs the 3GPP LTE/LTE-A cellular communications technique, which supports high bandwidth transmission and operates in a different frequency band from NB-IoT. We suppose that the eNB owns L radio resources (termed backhaul resources (BRs)) for the MTCG-to-eNB links, and the eNB has many more BRs than ARs, i.e., K < L. Each MTCG-toeNB link occupies w BRs of the eNB, where  $w \in \{1, \dots, W\}$ ,  $W \leq L$ . Meanwhile, the eNB is also responsible for supporting H2H communications. H HTC devices (HTCDs) coexist with MTCDs in the serving areas of the network, and interact with the eNB through the HTCD-to-eNB links. The HTCDto-eNB links and MTCG-to-eNB links share the common BRs of the eNB, which allocates v BRs for each HTCDto-eNB link, where  $v \in \{1, \ldots, V\}, V \leq L$ . Assume that each HTCD sending data packets also follows an independent Poisson process with a mean rate of  $\lambda_h$ , and the size of the HTC packets also follows an exponential distribution with a mean size of  $\rho_h$ . Each MTCG-to-eNB link and HTCDto-eNB link also can support  $l_h$  transmission modes with different transmission rates  $C_h(\cdot)$ . Let  $\{\{\theta_{h,i}\}_{i=1}^{l_h+1}\}$  be the set of SNR boundary points, where  $\theta_{h,l_h+1} = \infty$ . When SNR  $\xi \in [\theta_{h,i}, \theta_{h,i+1})$ , the transmission rate  $C_h(\xi)$  of transmission mode i is equal to  $\mu_{h,i}$ , where  $i \in \{1, \ldots, l_h\}$ .

To maximize network revenue, the network must decide the access strategy for each MTCD (i.e., through the eNB or an MTCG), and optimize radio resource allocation for each wireless link. Thus, we formulate the radio resource allocation model in the software-defined IoT network as an SMDP problem. In SDN, all the decision-making procedures are carried out in the SDN controller. When a new MTC or HTC data packet arrives, the SDN controller first evaluates the expected system gain and system expense based on the current status information of the eNB and MTCGs. Then, the SDN controller decides whether to accept or reject the data packets, which MTCGs or eNB should be selected for MTC traffic transmission, and how to allocate radio resources to each wireless link. In the SMDP framework, the decisions adopted by the SDN controller are called actions, and the moments when decisions are made are termed *decision epochs*. The action chosen is based on the current system state of the network, which includes the current traffic load on each MTCG and the eNB. To make the optimal decision for the reward model, the SDN controller needs to obtain the system reward for each action before making any decision. The longterm expected average reward per unit time of the network is considered as the optimal criterion for the SMDP.

## **B.** Problem Formulation

This subsection aims to formulate the considered optimized problem as an SMDP. The corresponding system states, actions based on each state, and the reward model are described as follows.

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1) System States: The system state S of the software-defined IoT network can be represented by the number of current wireless links in the network with different numbers of radio resources occupied, as well as an event occurred in the system, which could be either the arrival or departure of a data packet. The system state space S can be denoted as follows, i.e.,

$$\mathcal{S} = \{ s \mid s = (\boldsymbol{g}_1, \boldsymbol{g}_2, \dots, \boldsymbol{g}_N, \boldsymbol{s}_1, \boldsymbol{s}_2, \dots, \boldsymbol{s}_N, \boldsymbol{s}_m, \boldsymbol{s}_h, e) \},$$
(1)

where  $\boldsymbol{g}_i, \boldsymbol{s}_i, \boldsymbol{s}_m, \boldsymbol{s}_h, i \in \{1, \dots, N\}$  are defined as

$$g_i = (g_{i,1}, g_{i,2}, \dots, g_{i,U}), 
 s_i = (s_{i,1}, s_{i,2}, \dots, s_{i,W}), 
 s_m = (s_{m,1}, s_{m,2}, \dots, s_{m,U}), 
 s_h = (s_{h,1}, s_{h,2}, \dots, s_{h,V}).$$
(2)

The above symbols are explained in detail below

- $g_i$ : a vector of  $g_{i,u}$ , where  $u \in \{1,\ldots,U\}$ .  $g_{i,u}$ represents the number of wireless links between the MTCDs and the *i*th MTCG that occupy u ARs. The total number of allocated ARs of the *i*th MTCG should satisfy  $\sum_{u=1}^{U} (ug_{i,u}) \leq K;$
- $s_m$ : a vector of  $s_{m,u}$ , where  $u \in \{1, \ldots, U\}$ . Similarly,  $s_{m,u}$  represents the number of MTCD-to-eNB links that occupy u ARs. The total number of allocated ARs of the eNB should also be subjected to  $\sum_{u=1}^{U} (us_{m,u}) \leq K$ ;  $s_i$ : a vector of  $s_{i,w}$ , where  $w \in \{1, \dots, W\}$ .  $s_{i,w}$  is the
- number of the wireless links between the *i*th MTCG and the eNB that occupy w BRs;
- $s_h$ : a vector of  $s_{h,v}$ , where  $v \in \{1,\ldots,V\}$ .  $s_{h,v}$ indicates the number of HTCD-to-eNB links that occupy v BRs. Thus, the total number of allocated BRs of the eNB should satisfy  $\sum_{i=1}^{N} \left( \sum_{w=1}^{W} (ws_{i,w}) \right) +$  $\sum_{v=1}^{V} (vs_{h,v}) \leq L;$ e: An event in the event set  $\mathcal{E}$ , i.e.,  $e \in \mathcal{E}$ .

The event set  $\mathcal{E}$  is denoted by

$$\mathcal{E} = \{A_m, A_h\} \cup \mathcal{D}_1 \cup \mathcal{D}_2 \cup \ldots \cup \mathcal{D}_N \cup \mathcal{D}_m \cup \mathcal{D}_h, \quad (3)$$

where  $\mathcal{D}_i, \mathcal{D}_m, \mathcal{D}_h, i \in \{1, \dots, N\}$  are defined as

$$\mathcal{D}_{i} = \{ D_{i,u,w} | \ u \in \{1, ..., U\}, \ w \in \{1, ..., W\} \},$$

$$\mathcal{D}_{m} = \{ D_{m,u} | \ u \in \{1, ..., U\} \},$$

$$\mathcal{D}_{h} = \{ D_{h,v} | \ v \in \{1, ..., V\} \}.$$

$$(4)$$

Each event is detailed as follows

 $A_m$ : the network receives a data packet from an MTCD.  $A_m$  can be further denoted as  $A_m$  $(\mu_1, \mu_2, ..., \mu_N, \mu_{N+1})$ , where  $\{\mu_i\}_{i=1}^N$  represent the transmission rate of the wireless link between the MTCD and the *i*th MTCG;  $\mu_{N+1}$  is the transmission rate of the

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MTCD-to-eNB link; and  $\{\mu_i\}_{i=1}^{N+1} \in \{\{\mu_{m,i}\}_{i=1}^{l_m}\}$ . The transmission rate of each wireless link is evaluated by the SDN controller according to the SNR;

- A<sub>h</sub>: the network receives a data packet from a HTCD.
   A<sub>h</sub> can be further denoted as A<sub>h</sub> = μ<sub>h</sub>, where μ<sub>h</sub> is the transmission rate of the HTCD-to-eNB link, and μ<sub>h</sub> ∈ {{μ<sub>h,i</sub>}<sup>l<sub>h</sub></sup><sub>i=1</sub>};
- $D_i$ : an MTC data packet departs from the *i*th MTCG, where  $D_{i,u,w}$  denotes the departure of an MTC data packet that occupies *u* ARs and *w* BRs;
- $\mathcal{D}_m$ : an MTC data packet departs from the eNB, where  $D_{m,u}$  represents the departure of an MTC data packet that occupies u ARs;
- $\mathcal{D}_h$ : a HTC data packet departs from the eNB, where  $D_{h,v}$  denotes the departure of a HTC data packet that occupires v BRs.

2) Actions: When an event e occurs, the SDN controller excutes an action a(s) from the action set  $\mathcal{A}_s$  according to the current state s of the network, i.e., $a(s) \in \mathcal{A}_s$ . The action set can be described by

$$\mathcal{A}_{s} = \begin{cases} \{(-1,0,0), (i,u,w), (N+1,u,0)\}, & e = A_{m} \\ \{(-2,0,0), (N+2,0,v)\}, & e = A_{h} \\ \{(0,0,0)\}, & e \in \mathcal{E} \setminus \{A_{m},A_{h}\}, \end{cases}$$
(5)

where  $\mathcal{E} \setminus \{A_m, A_h\}$  means the relative complement of  $\{A_m, A_h\}$  in  $\mathcal{E}$ , i.e.,  $\mathcal{D}_1 \cup \mathcal{D}_2 \cup \ldots \cup \mathcal{D}_N \cup \mathcal{D}_m \cup \mathcal{D}_h$ . As can be seen from (5), action a(s) is represented by a triplet, where the first element denotes the transmission strategy of a data packet (reject, transmit via an MTCG or the eNB), and the second and third elements denote the allocation of ARs and BRs, respectively. The details of each action is given as follows

- When e = A<sub>m</sub>, three types of actions can be chosen from set A<sub>s</sub>, i.e., reject, transmitting through an MTCG or the eNB. When 'reject' is chosen, no radio resources would be allocated, and thus a(s) = (-1,0,0); When an MTC packet is transmitted to the *i*th MTCG with u ARs and w BRs allocated, action a(s) ∈ {(i, u, w)}, where i ∈ {1,...,N}, u ∈ {1,...,U}, and w ∈ {1,...,W}; Similarly, when an MTC packet is transmitted to the eN-B with u ARs allocated, action a(s) ∈ {(N + 1, u, 0)};
- When  $e = A_h$ , the HTC data packet can be either rejected or accepted by the network. The rejection of a HTC packet is denoted by a(s) = (-2, 0, 0), and the acceptance of a HTC packet with v BRs allocated is described by  $a_s \in \{(N+2, 0, v)\}, v \in \{1, \dots, V\}$ ; and
- When e ∈ E \ {A<sub>m</sub>, A<sub>h</sub>}, i.e., the transmission of an MTC or HTC data packet is completed, and no other action is required to be taken by the SDN controller except for updating the status information from the eNB and MTCGs, which is described by a(s) = (0,0,0).

3) Rewards: According to the system state s and the corresponding action a, the system reward can be given by

$$r(s,a) = k(s,a) - \omega(s,a), \tag{6}$$

where k(s, a) is the lump sum income of the system by taking action a in state s, and  $\omega(s, a)$  is the expected system cost. k(s, a) can be further defined as follows

$$k(s,a) = \begin{cases} E_2 - \beta t_h, & e = A_h, a(s) = (N+2,0,v) \\ E_1 - \beta t_m, & e = A_m, a(s) = (N+1,u,0) \\ E_0 - \beta t_g, & e = A_m, a(s) = (i,u,w) \\ -P_1, & e = A_m, a(s) = (-1,0,0) \\ -P_2, & e = A_h, a(s) = (-2,0,0) \\ 0, & e \in \mathcal{E} \setminus \{A_m, A_h\}, \\ a(s) = (0,0,0). \end{cases}$$
(7)

When a HTC data packet is accepted by the eNB, the network earns an income of  $E_2$ . Meanwhile, the HTC data packet also consumes  $\beta t_h$  resources for occupying v BRs of the eNB during transmission, where  $t_h$  denotes the transmission time and  $\beta$  refers to the price per unit time with the same measurement unit as the income. Therefore,  $k(s, a) = E_2 - \beta t_h$ , when  $e = A_h$  and a(s) = (N + 2, 0, v). Considering that the channel quality and the number of allocated resources have an impact on the transmission time,  $t_h$  can be rewritten as

$$t_h = \frac{\delta_h}{v\mu_h},\tag{8}$$

where  $\delta_h$  is the length of the HTC packet, which follows an exponential distribution with a mean size of  $\rho_h$ .

For MTC data packets, the lump sum income of the system can be derived similarly to HTC data packets. When a data packet is transmitted to the eNB, the network earns  $E_1$  and pays  $\beta t_m$  for the occupation of u ARs of the eNB. On the other hand, when an MTC packet is relayed by an MTCG, the network earns  $E_0$  and expends  $\beta t_g$  for allocating u ARs on the MTCD-to-MTCG link and w BRs on the MTCG-to-eNB link. In these cases,  $t_m$  and  $t_g$  can be rewritten as

$$t_m = \frac{\delta_m}{u\mu_{N+1}},$$
  

$$t_g = \frac{\delta_m}{u\mu_m} + \frac{\delta_m}{w\mu_q},$$
(9)

where  $\delta_m$  is the length of the MTC packet which follows an exponential distribution with a mean size of  $\rho_m$ ;  $\mu_m$  and  $\mu_g$  denote the transmission rates of the MTCD-to-MTCG link and MTCG-to-eNB link, respectively, which are decided by whichever MTCG is selected by the SDN controller according to the system revenue. When the MTC data packet is forwarded by the *i*th MTCG, we have  $\mu_m = \mu_i$ , and  $\mu_g$  can be derived by  $C_h(\xi)$  according to the SNR  $\xi$  of the wireless link between the *i*th MTCG and the eNB.

When an MTC or HTC data packet is rejected, the network receives a penalty of  $(-P_1)$  or  $(-P_2)$ , respectively. Moreover, the network receives no income when a data packet transmission is completed, i.e., k(s,a) = 0 for a(s) = (0,0,0) and  $e \in \mathcal{E} \setminus \{A_m, A_h\}$ .

On the other hand, the expected system cost  $\omega(s, a)$  given in (6) can be defined as

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$$\omega(s,a) = o(s,a)\tau(s,a), \quad a(s) \in \mathcal{A}_s \tag{10}$$

where  $\tau(s, a)$  is the expected time duration between two continuous decision epochs; o(s, a) is the cost rate of the system, which is determined by the total number of occupied ARs and BRs

$$o(s,a) = \sum_{i=1}^{N} \left( \sum_{u=1}^{U} ug_{i,u} + \sum_{w=1}^{W} ws_{i,w} \right) + \sum_{u=1}^{U} us_{m,u} + \sum_{v=1}^{V} vs_{h,v}.$$
(11)

#### III. SOLUTION TO THE SMDP PROBLEM

In this section, we first present the state transition probability that has a significant effect on deriving the optimal policy. Then, an average reward criterion is used as the performance criterion, since we focus primarily on the long term performance of the network. Our main objective is to maximize the expected average reward of the system by making optimal decisions at decision epochs. Finally, a relative value iteration algorithm is utilized to obtain the optimal policy.

#### A. Transition probability

An action taken by the SDN controller causes the state transition, which is characterized by the state transition probability. To obtain the transition probability, one should first derive the mean rate of events. When an action a is selected at the current state s, the system will transit to the next state j before the next decision epoch. The interval between two continuous decision epochs is denoted by  $\tau(s, a)$ , which is given in (10). Therefore, the mean rate  $\gamma(s, a)$  of events for a given s and a is the sum rate of all events in the system, which is the reciprocal of  $\tau(s, a)$ . To ease of exposition, the following two mathematical operators  $\mathbf{1}_{x,y}$  and  $\bar{\mathbf{1}}_{x,y}$  are introduced

$$\mathbf{1}_{x,y} = \begin{cases} 1, & \text{if } x = y \\ 0, & \text{if } x \neq y, \end{cases} \quad \bar{\mathbf{1}}_{x,y} = \begin{cases} 0, & \text{if } x = y \\ 1, & \text{if } x \neq y. \end{cases}$$
(12)

Accordingly, the mean rate  $\gamma(s, a)$  of events can be expressed as

$$\gamma(s,a) = \tau^{-1}(s,a)$$

$$= \begin{cases} \gamma_0(s,a), & e \in \mathcal{E} \setminus \{A_m, A_h\} \\ & e = A_m, a(s) = (-1,0,0) \\ & e = A_h, a(s) = (-2,0,0) \end{cases}$$

$$\gamma_0(s,a) + \frac{1}{\rho_h} v \bar{\mathcal{C}}_h, & e = A_h, a(s) = (N+2,0,v) \\ \gamma_0(s,a) + \frac{1}{\rho_m} u \bar{\mathcal{C}}_m, & e = A_m, a(s) = (N+1,u,0) \\ \gamma_0(s,a) & e = A_m, a(s) = (i,u,w), \\ + \frac{1}{\rho_m} \bar{\mathbf{1}}_{s'_{i,w},0} u \bar{\mathcal{C}}_m \\ + \frac{1}{\rho_m} \bar{\mathbf{1}}_{g'_{i,u},0} w \bar{\mathcal{C}}_g, \end{cases}$$
(13)

where  $\gamma_0(s, a)$  can be further denoted by

$$\begin{aligned} \gamma_{0}(s,a) &= M\lambda_{m} + H\lambda_{h} \\ &+ \frac{1}{\rho_{m}} \sum_{i=1}^{N} \sum_{u=1}^{U} \sum_{w=1}^{W} \left( \bar{\mathbf{1}}_{s_{i,w}^{\prime},0} g_{i,u} u \bar{\mathbf{C}}_{m} + \bar{\mathbf{1}}_{g_{i,u}^{\prime},0} s_{i,w} w \bar{\mathbf{C}}_{g} \right) \\ &+ \frac{1}{\rho_{m}} \sum_{u=1}^{U} s_{m,u} u \bar{\mathbf{C}}_{m} + \frac{1}{\rho_{h}} \sum_{v=1}^{V} s_{h,v} v \bar{\mathbf{C}}_{h}, \end{aligned}$$
(14)

where  $g_{i,u}$  and  $s_{i,w}$  are the elements in the current state s, which are given in (2);  $g'_{i,u}$  and  $s'_{i,w}$  are the corresponding values of  $g_{i,u}$  and  $s_{i,w}$  in the next state j, respectively, which are affected by the action of the current state a(s).

When a data packet transmission is completed, or a data packet is rejected by the network, the mean rate of events  $\gamma(s, a)$  is equal to  $\gamma_0(s, a)$  as shown in (14), where  $M\lambda_m$ and  $H\lambda_h$  are the total arrival rates of MTC and HTC data packets, respectively. Moreover, the reset part of  $\gamma_0(s, a)$  is the departure rate of the system. Since the long term performance of the network is mainly concerned, we take the average transmission rate as the capacity for each wireless link. Denote by  $\bar{C}_m$  the average transmission rate of the MTCD-to-MTCG and MTCD-to-eNB links;  $\bar{C}_g$  and  $\bar{C}_h$  the average transmission rates of the MTCG-to-eNB links and HTCD-to-eNB links, respectively.  $\bar{C}_m$  and  $\bar{C}_h$  are defined as follows [24]

$$\bar{\mathbf{C}}_m = \sum_{i=1}^{l_m} \mathbf{P}_m(i)\mu_{m,i},$$

$$\bar{\mathbf{C}}_h = \sum_{i=1}^{l_h} \mathbf{P}_h(i)\mu_{h,i},$$
(15)

where  $P_m(i)$  and  $P_h(i)$  are the probabilities of applying the transmission mode *i* of  $C_m(\cdot)$  and  $C_h(\cdot)$ , respectively. To simplify the analysis, it is assumed that each transmission mode is chosen with an equal probability, i.e.,  $P_m(i) = 1/l_m$  and  $P_h(i) = 1/l_h$ .  $\bar{C}_g$  is compute through averaging the transmission rates of the wireless links between N MTCGs and the eNB.

When a HTC data packet is accepted by the network, the eNB allocates v more radio resources for its transmission, and thus  $\gamma(s, a)$  is larger than  $\gamma_0(s, a)$  by  $\frac{1}{\rho_h}v\bar{C}_h$ . Similarly, the mean rate of events is equal to  $\gamma_0(s, a) + \frac{1}{\rho_m}u\bar{C}_m$ , when an MTC data packet is transmitted to the eNB with u resources occupied. For the MTC data packet forwarded by the MTCG, the mean rate of events is equal to  $\gamma_0(s, a) + \frac{1}{\rho_m}\bar{\mathbf{1}}_{s'_{i,w},0}u\bar{\mathbf{C}}_m + \frac{1}{\rho_m}\bar{\mathbf{1}}_{g'_{i,u},0}w\bar{\mathbf{C}}_g$ . The state transition probability is represented by q(j|s, a),

The state transition probability is represented by q(j|s, a), which means the state is transited from state s to state j under action a. To simplify the expression of q(j|s, a), the following function is defined

$$F_s(x_1 + m_1, x_2 + m_2, ..., x_n + m_n, e') \equiv (..., x_1 + m_1, ..., x_2 + m_2, ..., x_n + m_n, ..., e'),$$
(16)  
if  $s = (..., x_1, ..., x_2, ..., x_n, ..., e),$ 

which means that only element  $x_i$  in state s is updated to  $x_i + m_i$  and other elements remain the same, when the system enters the next state j with new event e'. Therefore, the state transitions for different actions can be obtained as follows

For  $e = A_m$  in the current state s, q(j|s,a) can be obtained via  $(18) \sim (19)$ , which are categorized by the available action a of the current state s. When a(s) =(-1, 0, 0), q(j|s, a) can be calculated by (19), while (17) and (18) are used for the cases of a = (N+1, u, 0) and a = (i, u, w), respectively. Each transition probability in each formula is calculated in the same way. That is, the mean rate of all events  $\gamma(s, a)$  divides the rate of the event e with action a. When the event e' in the next state j is  $A_m$ , coefficient  $1/l_m^{(N+1)}$  is introduced due to the assumption that  $\{\mu_i\}_{i=1}^{N+1}$  of  $A_m$  adopt each transmission mode with the same probability. For the same reason, the coefficient  $1/l_h$  is also used for cases where  $e' = A_h$ . In addition, the symbols in (18)~(19) are explained as:  $i, i_1 \in \{1, \ldots, N\}$  are the indexes of the MTCG;  $u, u_1 \in \{1, \ldots, U\}$  denote the number of ARs;  $w, w_1 \in \{1, \ldots, W\}$  represent the number of BRs for the MTCG-to-eNB link, and  $v, v_2 \in \{1, \ldots, V\}$  are the resource allocated for the HTCD-to-eNB link;

$$\begin{split} q(j|s,a) \mid_{a=(N+1,u,0)} &= \\ \begin{cases} \frac{1}{l_m^{(N+1)}} \cdot \frac{M\lambda_m}{\gamma(s,a)}, & j = \mathcal{F}_s(s_{m,u}+1,A_m) \\ \frac{1}{l_h} \cdot \frac{H\lambda_h}{\gamma(s,a)}, & j = \mathcal{F}_s(s_{m,u}+1,A_h) \\ \frac{g_{i,u_1}u_1\bar{\mathcal{C}}_m + s_{i,w}w\bar{\mathcal{C}}_g}{\rho_m\gamma(s,a)}, & j = \mathcal{F}_s(s_{m,u}+1,g_{i,u_1}-1, s_{i,w}-1,D_{i,u_1,w}) \\ \frac{(s_{m,u_1}+1_{u,u_1})u_1\bar{\mathcal{C}}_m}{\rho_m\gamma(s,a)}, & j = \mathcal{F}_s(s_{m,u}+1,s_{m,u_1}-1,D_{m,u_1}) \\ \frac{s_{h,v}v\bar{\mathcal{C}}_h}{\rho_h\gamma(s,a)}, & j = \mathcal{F}_s(s_{m,u}+1,s_{h,v}-1,D_{h,v}) \end{split}$$
(17)

For e = A<sub>h</sub> in the current state s, q(j|s, a) can also be obtained by (19) with a(s) = (-2, 0, 0). When a(s) = (N+2, 0, v), q(j|s, a) can be calculated by (20), which is derived similarly to (18)~(19). Moreover, each symbol in (20) is the same as in (18)~(19);

 $q(j|s,a)|_{a(s)=(N+2,0,v)} =$ 

$$\begin{cases} \frac{1}{l_m^{(N+1)}} \cdot \frac{M\lambda_m}{\gamma(s,a)}, & j = F_s(s_{h,v} + 1, A_m) \\ \frac{1}{l_h} \cdot \frac{H\lambda_h}{\gamma(s,a)}, & j = F_s(s_{h,v} + 1, A_h) \\ \frac{g_{i,u}u\bar{C}_m + s_{i,w}w\bar{C}_g}{\rho_m\gamma(s,a)}, & j = F_s(s_{h,v} + 1, g_{i,u} - 1, s_{i,w} - 1, D_{i,u,w}) \\ \frac{s_{m,u}u\bar{C}_m}{\rho_m\gamma(s,a)}, & j = F_s(s_{h,v} + 1, s_{m,u} - 1, D_{m,u}) \\ \frac{(s_{h,v_1} + \mathbf{1}_{v,v_1})v_1\bar{C}_h}{\rho_h\gamma(s,a)}, & j = F_s(s_{h,v} + 1, s_{h,v_1} - 1, D_{h,v_1}) \end{cases}$$
(20)

$$q(j|s,a)|_{a=(i,u,w)}$$

$$\frac{1}{l_{m}^{(N+1)}} \cdot \frac{M\lambda_{m}}{\gamma(s,a)}, \qquad j = F_{s}(g_{i,u} + 1, s_{i,w} + 1, A_{m}) \\
\frac{1}{l_{h}} \cdot \frac{H\lambda_{h}}{\gamma(s,a)}, \qquad j = F_{s}(g_{i,u} + 1, s_{i,w} + 1, A_{h}) \\
\frac{g_{i_{1},u_{1}}u_{1}\bar{C}_{m} + s_{i_{1},w_{1}}w_{1}\bar{C}_{g}}{\rho_{m}\gamma(s,a)}, \qquad j = F_{s}(g_{i,u} + 1, g_{i_{1},u_{1}} - 1, s_{i,w} + 1, s_{i_{1},w_{1}} - 1, D_{i_{1},u_{1},w_{1}}), \\
\frac{(g_{i,u_{1}} + 1_{u,u_{1}})u_{1}\bar{C}_{m}}{\rho_{m}\gamma(s,a)}, \qquad j = F_{s}(g_{i,u} + 1, g_{i,u_{1}} - 1, s_{i,w} + 1, s_{i,w_{1}} - 1, D_{i_{1},u_{1},w_{1}}), \\
\frac{(g_{i,u_{1}} + 1_{u,u_{1}})u_{1}\bar{C}_{m}}{\rho_{m}\gamma(s,a)}, \qquad j = F_{s}(g_{i,u} + 1, g_{i,u_{1}} - 1, s_{i,w_{1}} - 1, D_{i,u_{1},w_{1}}), \\
\frac{s_{m,u_{1}}u_{1}\bar{C}_{m}}{\rho_{m}\gamma(s,a)}, \qquad j = F_{s}(g_{i,u} + 1, s_{i,w} + 1, s_{m,u_{1}} - 1, D_{m,u_{1}}), \\
\frac{s_{h,v}v\bar{C}_{h}}{\rho_{h}\gamma(s,a)}, \qquad j = F_{s}(g_{i,u} + 1, s_{i,w} + 1, s_{m,u_{1}} - 1, D_{m,u_{1}}), \\
(18)$$

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$$q(j|s,a)|_{a(s)\in\{(-1,0,0),(-2,0,0),(0,0,0)\}} =$$

$$\begin{cases} \frac{1}{l_m^{(1+1)}} \cdot \frac{MA_m}{\gamma(s,a)}, & j = \mathcal{F}_s(A_m) \\ \frac{1}{l_h} \cdot \frac{H\lambda_h}{\gamma(s,a)}, & j = \mathcal{F}_s(A_h) \\ \frac{g_{i,u} \cdot u\bar{\mathcal{C}}_m + s_{i,w} \cdot w\bar{\mathcal{C}}_g}{\rho_m \gamma(s,a)}, & j = \mathcal{F}_s(g_{i,u} - 1, s_{i,w} - 1, D_{i,u,w}) \\ \frac{s_{m,u} u\bar{\mathcal{C}}_m}{\rho_m \gamma(s,a)}, & j = \mathcal{F}_s(s_{m,u} - 1, D_{m,u}) \\ \frac{s_{h,v} v\bar{\mathcal{C}}_h}{\rho_h \gamma(s,a)}, & j = \mathcal{F}_s(s_{h,v} - 1, D_{h,v}) \end{cases}$$

$$(19)$$

• For  $e \in \mathcal{E} \setminus \{A_m, A_h\}$  in the current state s, the action a(s) = (0, 0, 0) and q(j|s, a) can be also obtained by (19).

## B. Average reward model

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The time duration between two continuous decision epochs  $\tau(s, a)$  follows an exponential distribution, whose cumulative distribution function (CDF) is shown below

$$F(t|s,a) = 1 - e^{-\gamma(s,a)t}, \quad t > 0.$$
 (21)

Therefore, the expected average reward r(s, a) during time  $\tau(s, a)$  can be calculated based on the average reward model defined in [25], i.e.,

$$\begin{aligned} r(s,a) &= k(s,a) - o(s,a) E_s^a \{\tau\} \\ &= k(s,a) - \frac{o(s,a)}{\gamma(s,a)}. \end{aligned}$$
(22)

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## C. Solution

The average reward of policy  $\pi$  is defined as

$$\psi^{\pi} = \lim_{N \to \infty} \frac{E^{\pi} \left\{ \sum_{n=1}^{N} r(s_n, a_n) \right\}}{E^{\pi} \left\{ \sum_{n=1}^{N} \tau(s_n, a_n) \right\}},$$
(23)

where  $s_n$  and  $a_n$  refer to the state and action at the decision epoch n, respectively. The optimal reward  $\psi^*$  is denoted as

$$\psi^* = \psi^{\pi^*} = \sup_{\pi \in \Pi} \psi^{\pi},$$
 (24)

where  $\Pi$  is the set of any feasible policy  $\pi$ . Moreover,  $\pi^*$  is the optimal policy that can be derived by solving the Bellman equation, i.e.,

$$\nu(s) = \max_{a \in \mathcal{A}_s} \{ r(s, a) - \psi \tau(s, a) + \sum_{j \in \mathcal{S}} q(j|s, a) \nu(j) \}, \ s \in \mathcal{S}$$
(25)

where  $\nu(s)$  is the potential function of state s. Then, the uniformization transformation is applied to transform the average reward SMDP to a discrete-time model so as to simplify the analysis [25]. To realize uniformization, parameter  $\eta$  is introduced and defined as

$$\eta \equiv M\lambda_m + H\lambda_h + K(N+1)\frac{\mu_{m,l_m}}{\rho_m} + L \max\left\{\frac{\mu_g}{\rho_m}, \frac{\mu_{h,l_h}}{\rho_h}\right\}, \quad \eta < \infty.$$
(26)

Thus, the uniformed reward function  $\tilde{r}(s, a)$  and uniformed transition probability  $\tilde{q}(j|s, a)$  are obtained as

$$\widetilde{r}(s,a) = r(s,a)\frac{\gamma(s,a)}{\eta},$$
(27)

$$\widetilde{\psi} = \frac{\psi}{\eta},\tag{28}$$

$$\widetilde{q}(j|s,a) = \begin{cases} 1 - \frac{[1 - q(s|s,a)]\gamma(s,a)}{\eta}, & j = s\\ \frac{q(j|s,a)\gamma(s,a)}{\eta}, & j \neq s \end{cases}.$$
(29)

Therefore, the Bellman equiation can be rewritten as

$$\widetilde{\nu}(s) = \max_{a \in \mathcal{A}_s} \{ \widetilde{r}(s, a) - \widetilde{\psi} + \sum_{j \in \mathcal{S}} \widetilde{q}(j|s, a) \widetilde{\nu}(j) \}, \ s \in \mathcal{S}.$$
(30)

Since the proposed SMDP model has finite state and action spaces, a relative value iteration algorithm can be applied to solve the Bellman equation as shown in **Algorithm 1**, which often offers a much faster rate of convergence with respect to the span seminorm [25]. Let  $\Phi(\nu)$  be the span of vector  $\nu$ , which is defined as follows

$$\Phi(\boldsymbol{\nu}) = \max_{s \in \mathcal{S}} \nu(s) - \min_{s \in \mathcal{S}} \nu(s).$$
(31)

For  $\boldsymbol{\nu} \in \mathcal{V}$ ,  $\Phi(\boldsymbol{\nu})$  is a seminorm on  $\mathcal{V}$ . In the result, the relative value iteration algorithm can obtain a vector of decision rules  $d_{\varepsilon}(s)$  that constitute the optimal policy  $\pi^*$ .

Algorithm 1 Relative Value Iteration Algorithm	
Step 1. Initialization:	

- Select  $\widetilde{\boldsymbol{\nu}}^0 \in \mathcal{V}$ .
- Choose a base state  $s^* \in S$ , and specify  $\varepsilon > 0$ .
- Set  $w^0 = \tilde{\nu}^0 \tilde{\nu}^0(s^*)e$ , where e is a vector of ones.

• Set n = 0. Step 2. Set

$$egin{aligned} \widetilde{oldsymbol{
u}}^{n+1} &= \max_{a\in\mathcal{A}_s}\{oldsymbol{r}_a+\mathbf{P}_aoldsymbol{w}^n\}, \ oldsymbol{w}^{n+1} &= \widetilde{oldsymbol{
u}}^{n+1}-\widetilde{
u}^n(s^*)oldsymbol{e}. \end{aligned}$$

where  $r_a$  is the vector of  $\tilde{r}(s, a)$ ;  $\mathbf{P}_a$  is the transition probability matrix under action a, which consists of  $\tilde{q}(j|s, a)$ .

**Step 3.** If  $\Phi(\tilde{\nu}^{n+1} - \tilde{\nu}^n) < \varepsilon$ , go to **Step 4**. Otherwise, n = n + 1 and return to **Step 2**.

**Step 4.** Choose  $d_{\varepsilon} \in \underset{a \in \mathcal{A}_s}{\operatorname{arg\,max}} \{ \mathbf{r}_a + \mathbf{P}_a \boldsymbol{\nu}^n \}.$ 

#### IV. PERFORMANCE EVALUATION

To evaluate the performance of our proposed radio resource allocation scheme for the SDN-based IoT network, we develop a Matlab simulator. In this section, we first introduce the simulation environment and then analyze the numerical results.

In the simulation, we consider an SDN-based IoT network that consists of an eNB, an MTCG, and an SDN controller. There are also 200 MTCDs and 20 HTCDs evenly scattered in the serving area of the network. Each MTCG and the eNB contains up to K = 3 ARs, and the eNB also owns L = 5 BRs. The maximum number of ARs allocated to each MTCD-to-eNB and MTCD-to-MTCG wireless link is U = 2. That is, a wireless link can be assigned 1 or 2 units of ARs based on the decision taken by the SDN controller. Moreover, the eNB can allocate W = 2 and V = 2 BRs at most to each MTCG-to-eNB and HTCD-to-eNB link, respectively. We assume that the MTCD-to-MTCG and MTCD-to-eNB links support  $l_m = 2$  transmission modes with rates  $\mu_{m,1} = 10$ and  $\mu_{m,2} = 20$ . The MTCG-to-eNB and HTCD-to-eNB links also support  $l_h = 2$  transmission modes, which provide higher transmission rates, i.e.,  $\mu_{h,1} = 100$  and  $\mu_{h,2} = 200$ . In order to expedite the execution time and to reduce memory usage, the aforementioned system parameters are configured in a small scale, which helps reduce the number of states in the state space as well as the number of feasible actions corresponding to the states. This will not affect the generality and accuracy of our simulations.

Moreover, the average sizes of the MTC and HTC packets are  $\rho_m = 5$  and  $\rho_h = 50$ , respectively. Each HTCD transmits data packets to the network with an arrival rate of  $\lambda_h = 0.1$ . To further study the relationship among the parameters, the arrival rate of the MTC data packets  $\lambda_m$  varies from 0.001 to 0.1 for performance comparison, which is specified in small values due to infrequent M2M communications traffic. The other parameters used in the simulation are summarized in Table I.

TABLE I. SIMULATION PARAMETERS.

Parameter	Value	Parameter	Value
Ν	1	$\lambda_m$	0.001 - 0.1
M	200	$\lambda_h$	0.1
H	20	$\rho_m$	5
K	3	$\rho_h$	50
L	5	$E_0$	10
U	2	$E_1$	10
V	2	$E_2$	10
W	2	$P_1$	10
β	10	$P_2$	5
$\mu_{h,1}$	100	$\mu_{m,1}$	10
$\mu_{h,2}$	200	$\mu_m$ 2	20

For comparison purposes, we will also benchmark the proposed scheme against the following comparative schemes:

- **Greedy scheme:** The algorithm that aims at maximizing the system current reward at the decision epoch;
- Channel precedence (CP) scheme: a heuristic algorithm that is based on the channel quality of the wireless links between the MTCD and the candidate eNB or MTCGs. In general, the wireless link with a better channel quality can provides a higher transmission rate. The SDN controller chooses the eNB or MTCG that can provide the highest transmission rate with enough spare resources for MTC packet transmission, which is always allocated as much resources as possible; and
- **eNB precedence (EP) scheme:** When there are spare ARs available in the eNB, the SDN controller gives priority to selecting the eNB for MTC data packet transmission, which is always allocated with the maximal number of ARs that the eNB can support. Otherwise, the SDN controller makes the decision that maximizes the system current reward in the decision epoch.

Furthermore, several metrics are specified to evaluate the performance of our proposed scheme. The first metric is the blocking rate of data packets, which represents the probability of data packets rejected by the network. The second metric is the expected average reward of the network defined in (22).

Before compare the performance of our proposed scheme with those of the other reference schemes, we first present the action probabilities with different arrival rates of MTC data packets. Fig. 2 shows that the SDN controller is more likely to select the eNB for MTC data packet transmission when  $\lambda_m$  is small. This is because the transmission via the eNB is through only one wireless link, which may consume less transmission time and receive a good system reward. With the increase of  $\lambda_m$ , the eNB allocates more ARs for MTC data packet transmission, which results in insufficient ARs in the eNB. Thus, the action probability of selecting the eNB decreases, while the action probability of selecting the MTCG increases. When an MTC data packet is transmitted by the eNB, the probabilities of resource allocation under each possible action (i.e.,  $a \in \{(2, 1, 0), (2, 2, 0)\}$ ) are depicted in Fig. 3. When  $\lambda_m$ 



Fig. 2. Action probabilities of selecting the MTCG and eNB with various arrival rates of MTC data packets per MTCD.



Fig. 3. Probabilities of resource allocation when selecting the eNB for MTC data packet transmission with various arrival rates of MTC data packets per MTCD.

is small, the eNB tends to allocate two ARs for the MTCDto-eNB links, while one AR is more likely to be allocated to the MTCD-to-eNB links with the increase of  $\lambda_m$ . This is because there are sufficient ARs in the eNB, which allocates the most number of ARs to maximize the system reward with a high probability. When an MTC data packet is forwarded via the MTCG, the probabilities of resource allocation to the MTCD-to-MTCG and MTCG-to-eNB links with various actions are shown in Fig. 4. With the increase of  $\lambda_m$ , almost all the curves increase except for the one that allocates two ARs to the MTCD-to-MTCG links, which starts decreasing slightly around the point where  $\lambda_m = 0.07$ . Moreover, it can be seen that the MTCD-to-MTCG and MTCG-to-eNB links are allocated two BRs and two ARs with high probabilities, respectively. This is because the network load is not high enough within the range of  $\lambda_m \in [0.001, 0.1]$ , and the MTCG

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Fig. 4. Probabilities of resource allocation when selecting the MTCG for MTC data packet transmission with various arrival rates of MTC data packets per MTCD.



Fig. 5. Blocking rate of MTC data packets with various arrival rates of MTC data packets per MTCD of each scheme.

still has resources available to accommodate new connections.

Then, we compare the blocking rates of the proposed SMDP-based model and other reference schemes. Fig. 5 plots the blocking rate of MTC data packets with various arrival rates of MTC data packets. When  $\lambda_m$  is very small, the blocking rates of MTC data packets of all the comparative schemes are close to zero. This is because there are sufficient ARs and BRs in the MTCG and eNB to be allocated for the wireless links, and the network seldom rejects MTC data packets. With the increase of  $\lambda_m$ , the MTCG and eNB have to allocate ARs and BRs to more wireless links so as to support more MTC data packet transmissions, which consumes available network resources resulting in a gradually increased blocking rate of MTC data packets. Among all the comparative schemes, the EP scheme has the highest blocking rate, since



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Fig. 6. Blocking rate of HTC data packets with various arrival rates of MTC data packets per MTCD of each scheme.

the EP scheme always first chooses the eNB for MTC data packet transmission, and then allocates ARs to the transmission as much as possible. This may exhaust the network resources rapidly resulting in insufficient spare resources for new arriving packets. For the same reason, the EP scheme also offers an inferior performance in terms of the MTC blocking rate. Compared to these two schemes, the proposed SMDP scheme and the CP scheme have better performances. The MTC blocking rate curves of the SMDP scheme and CP scheme cross at the point around  $\lambda_m = 0.07$ . When  $\lambda_m$  is smaller than the crossing point, the CP scheme has a lower blocking rate than the SMDP scheme under the low network load. This is attributable to the assumption that each wireless link adopts different transmission rates with the same probability, which makes the CP scheme select the MTCG or eNB equally likely. Therefore, the CP scheme has an effect on balancing the network load, and can reduce the possibility of network congestion. However, when  $\lambda_m$  is larger than the crossing point, our proposed SMDP scheme outperforms not only the CP scheme, but also all the other reference schemes. Moreover, the advantage of the SMDP scheme becomes more evident under a heavier traffic load (i.e, with the continuous increase of  $\lambda_m$ ). The reason is that the SMDP scheme aims to optimize resource allocation based on the long-term average reward, which tries to avoid the rejection of MTC data packets that may penalize the system reward.

On the other hand, the blocking rates of HTC data packets with various arrival rates of MTC data packets are plotted in Fig. 6. The blocking rate of HTC data packets of each scheme varies within a small range, which has little impact on the service quality of HTC users. Therefore, the total blocking rate of data packets has the same trend as that of MTC data packets as can be observed from Fig. 7. Our proposed SMDP scheme has an acceptable performance in terms of the blocking rate, and outperforms nearly all the other reference schemes except the CP scheme when  $\lambda_m$  is small.

We further compare the expected average rewards of the



Fig. 7. Total blocking rate of data packets with various arrival rates of MTC data packets per MTCD of each scheme.



Fig. 8. Average system reward with various arrival rates of MTC data packets per MTCD of each scheme ( $\lambda_h=0.1$ ).

SMDP-based model and the other reference schemes. The expected average rewards of the network with various arrival rates of MTC data packets are shown in Fig. 8. When  $\lambda_m$ is very low, the MTCG and eNB have a large umber of spare ARs and BRs available for MTCD access, and almost all MTC data packet transmissions are admitted by the network. Therefore, the expected average reward of each scheme increases rapidly with the increase of  $\lambda_m$ . However, when the resources of the MTCG and eNB are nearly depleted (i.e., aroud  $\lambda_m = 0.01$ ), the expected average rewards of the comparative schemes reach their peak value. However, when  $\lambda_m$  continues to increase, the expected average rewards of all the schemes decrease gradually. This is because the network suffers from a heavier traffic load and becomes more likely to reject MTC data packet transmission, which tends to penalize the system reward. Thus, the average reward of the network is a concave function of the arrival rate of MTC data packets. As

can be seen from Fig. 8, our proposed scheme significantly outperforms the other reference schemes. When  $\lambda_m$  is less than 0.04, there is no much difference between the SMDP scheme and the greedy scheme, because both schemes tend to adopt the same action. That is, when there are enough resources in the MTCG and eNB, the network tends to allocate resources to the wireless links as much as possible with the objective of achieving a good revenue. However, with the increase of  $\lambda_m$ , the difference between the SMDP scheme and the greedy scheme becomes larger, and the advantage of the SMDP scheme becomes more evident. This is because that the greedy scheme always allocates the maximum number of radio resources to MTC data packet transmission so as to achieve the highest system reward in the decision epoch, without consideration of whether the residual resources are enough for the next request. Thus, the greedy scheme may rapidly exhaust network resources, and the network may risk rejecting new data packets without sufficient radio resources. By contrast, the proposed SMDP-based radio resource allocation model makes decisions by considering both the instant lump sum income and the system expenses. From the long-term perspective, the SMDP scheme can achieve the optimized average reward. The CP and EP schemes do not aim to optimize the system reward. Therefore, they obtain worse expected average rewards.

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## V. CONCLUSIONS AND FUTURE WORK

In this paper, we investigated the resource allocation problem in the SDN-based IoT network. The problem was formulated as an SMDP process with the objective of maximizing the expected average reward of the network. Then, we obtained the optimal solution to the SMDP problem via a relative value iteration algorithm, which determines which MTCGs or eNB should be selected for MTC packet transmission and how many resources to be allocated for each wireless link. Simulation and numerical results were presented to demonstrate the superiority of our proposed scheme in comparison to the reference schemes.

In our future work, we will analyze the optimal resource allocation policy in the SDN-based IoT network that employs multiple M2M communications technologies to further validate the scalability of our proposed scheme.

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